

An asynchronous synchronous machine to control a flexible grid operation

Igor S. Suprunov^{1,*}, Dmitry V. Dvorkin^{1,*}, Aleksander N. Novikov² and Nikolay L. Novikov²

¹JSC “STC UPS Power system development”, Department of System Development, 109074, Moscow ,Kitaygorodskiy pass., 7, b.3. Russia.

²PJSC «FGC UES», 117630, Moscow, Akademika Chelomeya str., b.5A; «MPEI», 111250, Moscow, Krasnokazarmennaya 14; «JIHT RAS», 125412, Moscow, Izhorskaya st. 13 Bd.2. Russia.

Abstract. In this paper, a system of two asynchronous synchronous machines is proposed to control a power-flow in a meshed grid of 220/500 kV and a control algorithm, including a system parameterization guide. The proposed system is supposed to qualitatively affect the parameters of steady-state modes and short-circuit-caused transient processes and to increase the transfer capacity of the grid. Wherein, this double-fed synchronous machines complex allows for a significant increase in the power quality and reliability within the grid’s normal operation. Long-term dynamics is considered. As a result, it is shown that the proposed system parametrization provides a significant decrease in active power fluctuations and power system swinging.

1 Introduction

The interconnection of large *Power Grids (PG)* causes significant power flows through the tie-lines caused by a wide range of reasons, such as ensuring system reliability, time zone effect, etc. [1-3]. At the same time, the comparative chaotic nature of the large consumption centers development within the *PGs* and various elements of the tie-line connections leads to the fact that at a certain moment the existing grid infrastructure becomes incapable of ensuring the normal grid operation with the given actual interconnection tie flows due to insufficient power transmission capacity [1-3, 7-9,11-14,20-24]. One of the ways to increase the power transmission capacity is the application of flexible grid technologies based on the **Double-Fed Synchronous Machines Complex (DFSMC)** [3,8, 9,10,17,18]. This paper discusses the possibility of *DFSMC* application to increase the transmission capacity of the power grid and addresses the main technical and economic factors.

Since the flexible tie-line construction based on *DFSMC* requires high costs, therefore, for economic reasons, a combined solution is considered when *DFSMC* is installed in parallel with an autotransformer (*AT*) of comparable capacity instead of a traditional substation (*S*) with two *ATs* [17,30].

It is important to note that the contributions devoted to the *DFSMC* implementation for power transmission capacity increase have been previously published. So, in [32] it was proposed to use a current-limiting reactor connected in parallel with the *DFSMC*. The qualitative difference between the present paper and the aforementioned studies is that the latter examined the

properties of flexible interconnection tie-line in steady-state operation modes using simplified *PG* models. At the same time, the power generation equipment influence on the *PG* operation modes, as well as the voltage level impact on the load static characteristics were not taken into account. Thus, the issues of the operation modes comprehensive study of such a two-component device in a complex power system in steady-state and transient electromechanical processes remained unresolved.

1.1 Generalized model of DFSMC

To create a flexible interconnection tie-line, a technical complex based on *DFSMC* was chosen since it allows [3,4,8,11,12,25]:

- controlling the active power flow over a wide range;
- regulating the voltage and reactive power balance;
- providing complete electrical isolation of two large grid areas, limiting the short-circuit currents in-feeds and power quality distortions significantly;
- operating as an electric energy storage device.

The *DFSMC* is a controlled power device that includes two asynchronous synchronous machines (*ASM*), the rotors of which are united by a common shaft, connected to a three-phase power source, static frequency converters supplying the *DFSMC* excitation, a field regulator and other auxiliary systems (Fig. 1) [14,24 ,25,27-37]. On the rotor of each of the machines, there are two symmetrically positioned field coils *d-d* and *q-q*, which are powered from *P1* and *P2* by the alternating current with an ω frequency.

The rotation frequency of the rotor winding field ω_{ξ} is formed depending on the *DFSMC* shaft ω_r rotation speed and the orientation frequency $\omega_{r\xi}$, selected as the required frequency to ensure the *EMF* vector rotation E_{ξ}

* Corresponding authors: suprunov-is@ntc-msk.ru, dvorkin-dv@so-ups.ru

of each of the **DFSMC** machines (where $\xi=1,2$ corresponds to the subsystems numbers to which **DFSMC** machines are connected) [39]:

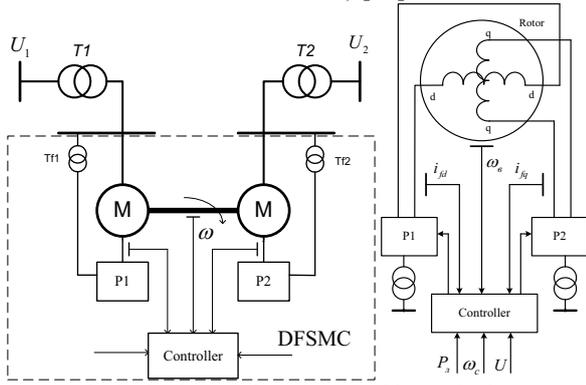


Fig.1. Principal scheme of **DFSMC** and its excitation system

$$\omega_{\xi} = \omega_{r\xi} - \omega_r. \quad (1)$$

In steady-state, the **EMF** rotation frequency E_{ξ} is determined by the expression:

$$\omega_{E\xi} = \omega_r + \omega_{\xi}. \quad (2)$$

Substituting expression (2) into (1), it can be seen that the **EMF** speed is determined according to the following law:

$$\omega_{E\xi} = \omega_{r\xi}. \quad (3)$$

Thus, the **EMF** speed is independent of the shaft speed ω_r . The frequency $\omega_{r\xi}$ is taken equal to the frequency ω_{ξ} of the subsystem ξ , electrically coupled with the machine. By changing the module and the **EMF** E_{ξ} phase by an automatic regulator, it is possible to implement the required **DFSMC** operation mode.

Since the magnetic flux in the rotor is formed as a result of two fluxes d - d and q - q , which are created by the excitation currents i_{fd} and i_{fq} , the rotor current corresponding to the resulting flux can be considered as a space vector in the rotor's own coordinate system (d , q), the coordinate axis of which (d axis) coincides with the d - d winding axis, and the abscissa axis (q axis) - with the q - q winding axis. It is assumed that the d -axis is imaginary and the q -axis is real. Then, the current in the field coil is determined according to the expression:

$$\dot{i}_f = i_{fq} + j \cdot i_{fd}. \quad (4)$$

In the given coordinates, the Park-Gorev equations take the following form:

$$\left. \begin{aligned} -\dot{u}_s &= \frac{1}{T} x_s \cdot \dot{i}_s + (p + j\omega_{r\xi}) \cdot (x_s \cdot \dot{i}_s + \dot{e}_f) \\ \dot{u}_f &= \dot{e}_f + [p + j(\omega_{r\xi} - \omega_r)] \cdot (\mu \cdot x_s \dot{i}_s + \dot{e}_f) \cdot T_f \\ T_f \cdot p\omega_r &= - \sum_{\xi=1}^2 \text{Re}(j \cdot e_{f\xi} \hat{i}_{s\xi}) \end{aligned} \right\} (5)$$

where $\dot{u}_f = x_{af} \cdot \dot{u}_f^{exc} / r_f$ is the reduced excitation voltage proportional to the voltage \dot{u}_f^{exc} , applied to the excitation field; $\dot{e}_f = x_{af} \cdot \dot{i}_f$ is the reduced rotor current proportional to the current \dot{i}_f ; $T_f = x_f / r_f$ is the time constant of the rotor winding; $\mu = x_{af}^2 / (x_s \cdot x_f)$ is the magnetic coupling ratio between the stator and rotor windings; $T = x_s / r_s$ is the time constant of the stator winding.

The automatic field controller **DFSMC** provides harmonic variations of the control voltage \dot{u}_f supplied to the field coils with the frequency ω_f , i.e. changing its module and phase depending on the operation mode of the combined control device and adjacent subsystems. To ensure the vector **EMF** rotation \dot{E}_{ξ} the excitation voltage $\dot{u}_{f\xi}$ is analytically determined according to the expression [25,39,41-44]:

$$\dot{u}_{f\xi}^{(or)} = u_{f\xi}'^{(or)} + ju_{f\xi}''^{(or)}, \quad (6)$$

or graphically as shown in Fig.2.

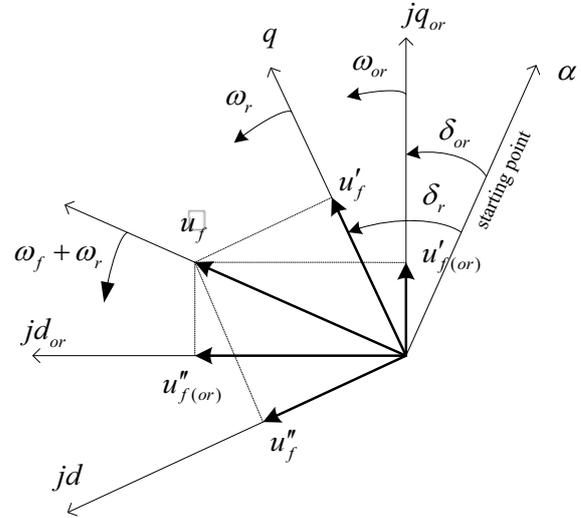


Fig.2. Vector \dot{u}_f and its projections on d_{or} , q_{or} and d_{ξ} , q_{ξ} axis

A change in the module and phase of **EMF** \dot{E}_{ξ} is caused by a change in the excitation voltage $\dot{u}_{fy\xi}$, according to the following law:

$$\left. \begin{aligned} u_{y\xi}'^{(or)} &= \alpha_0 + \sum_i W_i(p) \cdot \Delta P_i \\ u_{y\xi}''^{(or)} &= \beta_0 + \sum_j W_j(p) \cdot \Delta P_j \end{aligned} \right\} (7)$$

where $u_{y\xi}'^{(or)}$ and $u_{y\xi}''^{(or)}$ are the control voltage projections on the q_{or} and jd_{or} , respectively; α_0 and β_0 are the steady state parameters; $W_i(p)$, $W_j(p)$ are the transfer functions of the **automatic excitation controller (AEC)** of **ASM**; P_i , P_j are the parameters of the current operation mode.

The law or algorithm for the frequency ω_f calculation as a frequency function ω_{or} of the selected orientation vector and the converter shaft ω_r rotation frequency is uniquely determined from the condition of ensuring the steady state mode given the shaft frequency ω_r different from the adjacent subsystem frequency ω_{or} ($\omega_f = \omega_{or} - \omega_r$).

To increase the power transmitted over the interconnection tie-line, the case of a **DFSMC** with **AT** connected in parallel was considered [30,32,36]. For the simplicity reasons the **combined power flow controller** with **DFSMC** and **AT** is referred to as **CPFC** (Fig.3). Hereinafter, a combined device **DFSMC** with a power of 200 MW and an **AT** of 500/220 kV with a power of 500 MVA with complex impedance of 60 Ohm is considered.

The change in the flow P_L^A is carried out by regulating the active power flow through the **DFSMC**, P_y . In this case, the active power flowing through the **AT** is determined according to the expression [32,36]:

$$P_{AT} = P_L^A - P_y. \quad (8)$$

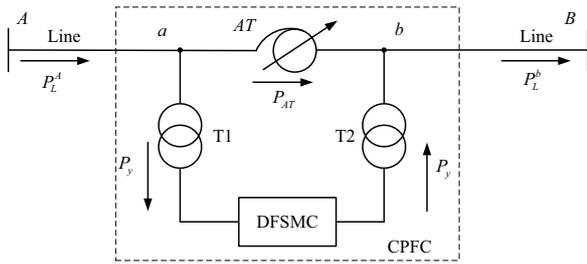


Fig.3. Principal scheme of an over-head-line (**OHL**) with **CPFC**

If the active power flow P_y , passing through the **DFSMC** decreases, then the power flowing through the **OHL** P_L^A decreases as well. A decrease in the power P_L^A flowing through the **OHL** also occurs with opposite direction of the flow P_y .

2 CPFC to control a flexible interconnection and to increase its transmission capacity

A schematic diagram of the studied inhomogeneous interconnection tie-line, containing parallel **OHLs** of 220 and 500 kV, connected at the intermediate substation through the **AT**, is shown in Fig.5. It is correct to characterize such a connection as heterogeneous and meshed. With an increase in active power transmitted from one system to another, the power flow in 220 kV **OHL** (lines 2 and 3) increases more rather than in 500 kV **OHL** (line 1). In this case, in certain operation modes, the transmission capacity limit of lines 2 and 3 can be reached, while the 500 kV line 1 will remain underloaded. Therefore, there is a need to limit the power flow along 220 kV lines depending on the operation mode in order to ensure the condition $P_{OHL220} \leq P_{limit}$.

For simplicity reasons, when considering the principle of the **OHL** power flow control using the proposed device, the assumption is introduced that the voltage vectors U_A and U_B difference at the terminals of the **OHL** in points *A* and *B* of the power grid is constant (Fig.4), that is [36]:

$$\delta_{AB} = \delta_{Aa} + \delta_{ab} + \delta_{bB} = const. \quad (9)$$

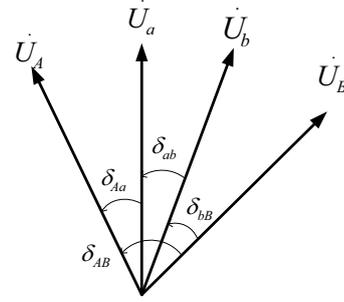


Fig.4. Vector diagram of the flexible transmission line

Due to the negligible values of power transmission sections' ohmic resistance, taking as an assumption that the voltage modules in points *a* and *b* are kept constant by regulating the output of synchronous machines' reactive power, it is fair to conclude that the transmitted active power along the sections P_L^A and P_L^b is determined by the angles between the voltage vectors U_A and U_a , U_b and U_B , that is, the angles δ_{Aa} and δ_{bB} and the modules of these voltages:

$$P_L^A = P_{max}^A \sin(\delta_{Aa} - \alpha_{Aa}); \quad (10)$$

$$P_L^b = P_{max}^b \sin(\delta_{bB} - \alpha_{bB}), \quad (11)$$

where P_{max}^A, P_{max}^b are the amplitude values of active power flowing in sections *A-a* and *b-B*.

Within these assumptions, the active power through the **AT** is determined by the expression:

$$P_{AT} = \frac{U_a U_b}{x_{AT}} \sin \delta_{ab}. \quad (12)$$

Based on the expression (12), the angle between the voltages U_a and U_b results in:

$$\delta_{ab} = \arcsin \frac{x_{AT} P_{AT}}{U_a U_b}. \quad (13)$$

The analysis of the expressions in Eqs.(10)-(13) shows that at $P_y > 0$ with a decrease in this power, the power P_{AT} and the angle δ_{ab} increase. Then, according to Eq.(9), the angles δ_{Aa} and δ_{bB} , the active power in the transmission sections P_L^A and P_L^b and the power transmitted along the line P_L decrease. By contrast, with increasing power $P_y > 0$ the described values increase as well.

An increase in the active power flow causes a voltage decrease at the substations, and in its turn, an increase in the reactive power flow leads to the line 2 electric current increase above the permissible value. The power-angle diagram of the combined device is shown, taking into account operation mode restrictions (for example, determined by the balance situation in the **PG**).

To prevent the monitored lines' overloading, it is necessary to reduce the active power flow. As a result, maintaining of constant active power-flow mode transforms into maintaining a given value of electrical current in the transmission line under consideration.

Fig.6 shows that up to an active power flow of 200 MW (section 1), no operation mode restrictions are required. At point 1, a specified power flow regulation is introduced up to the point 2, determined by the need to maintain this power flow value following the balance situation in the electric power system. Within section 2-

3, line 2 electric current limit violation occurs, caused by an increase in reactive power flow. In this case, it is necessary to reduce the transmitted active power below P_{L2_lim} . With a further increase in the power transmitted through the power grid, an unacceptable voltage drop occurs at the adjacent grid's substations. To prevent this drop, the power flow along line 2 must be further reduced. In this case, the increase in active power flow

through the grid decreases down to zero, after which the maximum allowed power flow continues to decrease.

Eventually, the control algorithm within the logic described in Eqs.(10)-(13) using **CPFC** allows to influence the power flow distribution, relieving the weak sections of the transmission grid. At the same time, this algorithm is efficient both for the example of the generalized system in Fig.3, and of the meshed grid presented in Fig.6.

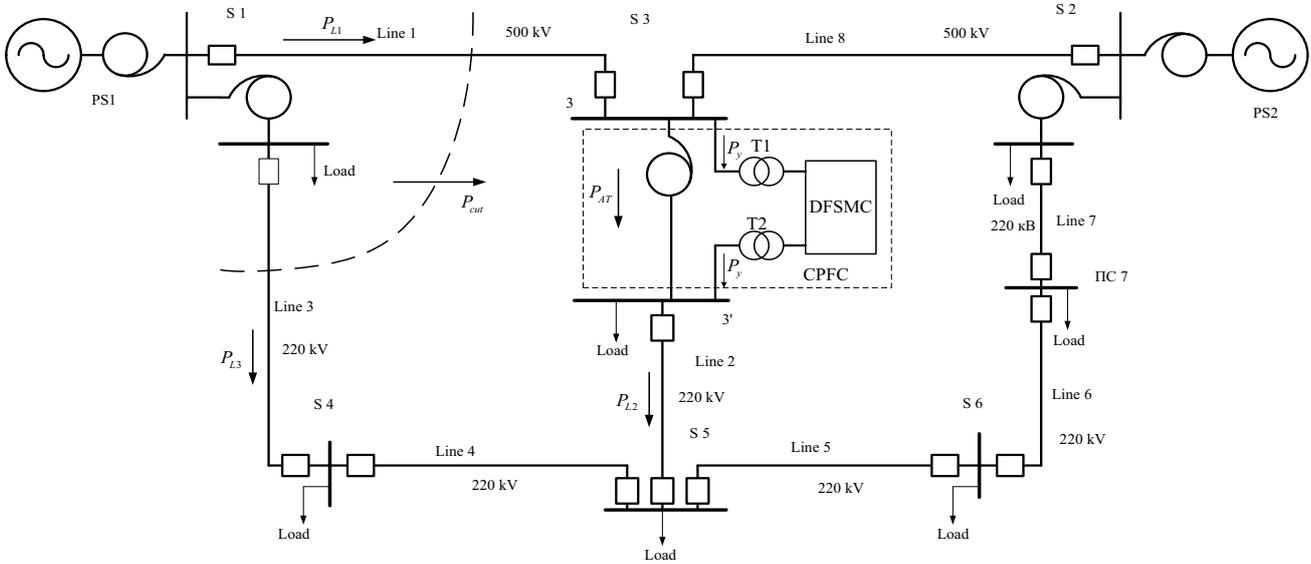


Fig.5. Principal scheme of the meshed grid with **CPFC**

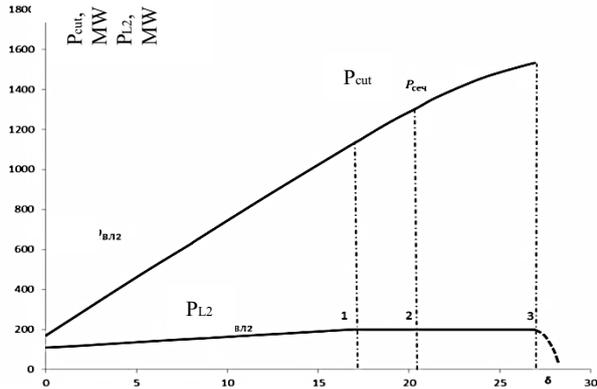


Fig.6. Power-angle diagram of the flexible with **CPFC**

3 Mathematical model of the meshed grid with CPFC to simulate transient processes

Fig.7 shows a diagram of the testing meshed grid used to calculate electromechanical transients.

The power grid consists of transformers, autotransformers, power transmission lines and loads. The equations for these elements are written in the d and q coordinate system of the backbone machine.

The equations describing each **ASM** in the d and q axes [18,25,31,32,39], result in:

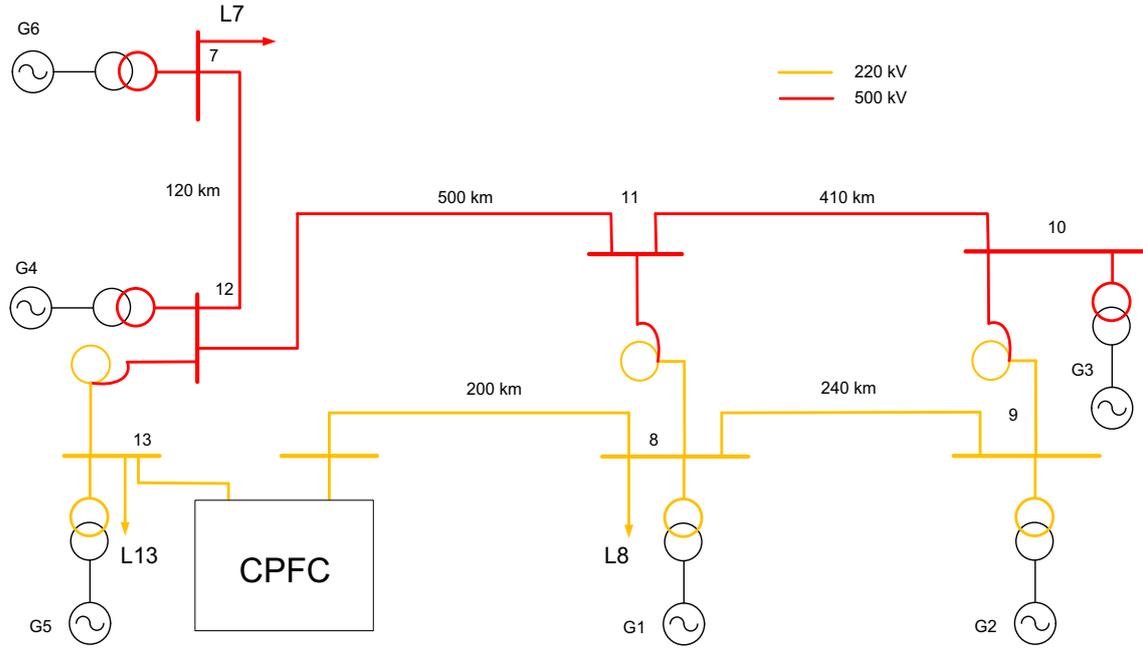


Fig.7. Grid model

$$U_d = \omega_s(E_d + i_q \cdot x_q), \quad (14)$$

$$U_q = \omega_s(E_q + i_d \cdot x_d), \quad (15)$$

$$p \cdot E'_d = \frac{1}{T_{q0eq}} [E_{de} - E_d] - s \cdot E'_d, \quad (16)$$

$$p \cdot E'_q = \frac{1}{T_{d0eq}} [E_{qe} - E_q] - s \cdot E'_q, \quad (17)$$

$$ps = \frac{1}{T_j} \cdot (P_e - P_T), \quad (18)$$

where U_d and U_q are the voltage vectors projections at the **ASM** connection points to the power system on the d and q axes; s is the **DFSMC** shaft slip relative to the synchronous speed; ω_s is the synchronous rotation speed of the power grid; E'_d and E'_q are the transient **EMFs** along the d and q axes, respectively; E_d and E_q are the idle **EMFs** along the d and q axes, respectively; i_d and i_q are the machine stator currents along the d and q axes, respectively; x_d and x_q are synchronous inductive reactance along the d and q axes (an assumption is made about the magnetic symmetry of the **ASM** rotor, $x_d = x_q = x$); P_e and P_T are, respectively, the machine power output supplied to the grid, and the mechanical power on the machine shaft; T_{d0eq} and T_{q0eq} are rotor's circuit time constants; T_j is rotor's inertia constant.

This equation system is a general case of a synchronous machine's mathematical model.

ASM G7 and **G8** are equipped with **AEC**. The excitation is controlled by the voltage deviation at the generator terminals from the set value and its time derivative, the **DFSMC** shaft slippage deviation from the set value, and its time derivative, as well as by the deviation of the control power P_c from the set value [25].

Forced **EMF** vector is decomposed in synchronous axes into E_{qe} and E_{de} (Fig.8).

In the presented vector diagram, δ_e is the angle shift of the subsystem generators' rotors, relative to the synchronous axis, which is determined by the following expression:

$$\delta_e = \frac{\sum_{i=1}^N T_{ji} \cdot \delta_i}{\sum_{i=1}^N T_{ji}}, \quad (19)$$

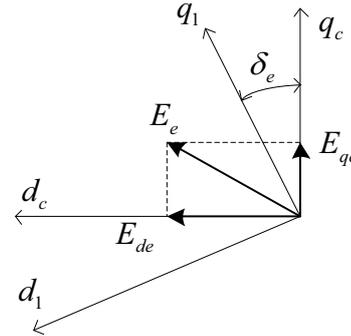


Fig.8. Forced EMF and its projections on axes d and q

In Eq.(19), N is the number of generators in the subsystem; T_{ji} is the inertia constant of the i -th generator; δ_i is the rotor's shift angle of the i -th generator relative to the synchronous axis.

If the vector decomposition of the idle **EMF** in synchronous coordinates has the values E_q and E_d , then, in the coordinates of the subsystem ξ , to which the **ASM** is connected, it would be $E_{q\xi}$ and $E_{d\xi}$, respectively (Fig.9).

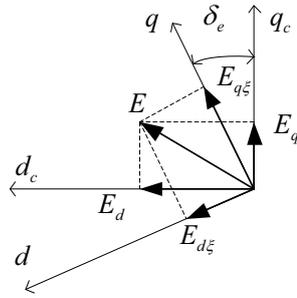


Fig.9. Idle EMF and its projections on axes d and q

From Fig.9 the following formulas are obtained (separately for each subsystem) [25,39]:

$$\left. \begin{aligned} E_{q\xi} &= E_q \cos(\delta_e) + E_d \sin(\delta_e) \\ E_{d\xi} &= -E_q \sin(\delta_e) + E_d \cos(\delta_e) \end{aligned} \right\} \quad (20)$$

The **ASM** excitation regulation law in general (in the active power maintenance operation mode) is given by:

$$V_{yd\xi} = E_{d\xi0} + V_s(\Delta s) + V_P(\Delta P), \quad (21)$$

$$V_{yq\xi} = E_{q\xi0} + V_U(\Delta U), \quad (22)$$

where V_s is the control law for shaft slippage deviation; Δs is the deviation of the shaft slip s ; V_P is the control law for the deviation of the transmitted power; ΔP is the deviation of the transmitted power P ; V_U is the control law for voltage deviation at the terminals of the

combined device; ΔU is the deviation of the voltage U ; $V_{yd\xi}$ and $V_{yq\xi}$ are the **ASM** excitation voltages in the projection on the d , q axes, rotating at the machine rotor speed; $E_{d\xi0}$ and $E_{q\xi0}$ are the initial values of the idle **EMF** of **ASM** in the projection on the d , q axes, rotating at the machine rotor speed.

Using (20)-(22) the reverse transition to synchronous axes is performed, taking into account the magnetic inertia of the rotor circuit:

$$\left. \begin{aligned} E_{qe\xi} &= (1 + k_f) \left(\begin{aligned} &V_{yq\xi} \cos(\delta_{e\xi}) - \\ &-V_{yd\xi} \sin(\delta_{e\xi}) \end{aligned} \right) - k_f E_{d\xi} \\ E_{de\xi} &= (1 + k_f) \left(\begin{aligned} &V_{yd\xi} \sin(\delta_{e\xi}) - \\ &-V_{yq\xi} \cos(\delta_{e\xi}) \end{aligned} \right) - k_f E_{q\xi} \end{aligned} \right\} \quad (23)$$

where k_f is the **ASM** rotor current feedback factor.

A three-phase short circuit (**SC**) at 500 kV line in the deficient part of the power system (subsystem 1) in 0.3 seconds at node 12 (Fig.7) was considered as a normative disturbance. The generators' rotors angles, the rotational speed of the generators shafts, as well as the generators active power outputs were considered as comparative values. In Fig.10-Fig.15 the changes of the indicated values are presented for the generators **G4** and **G5** closest to the **SC** point.

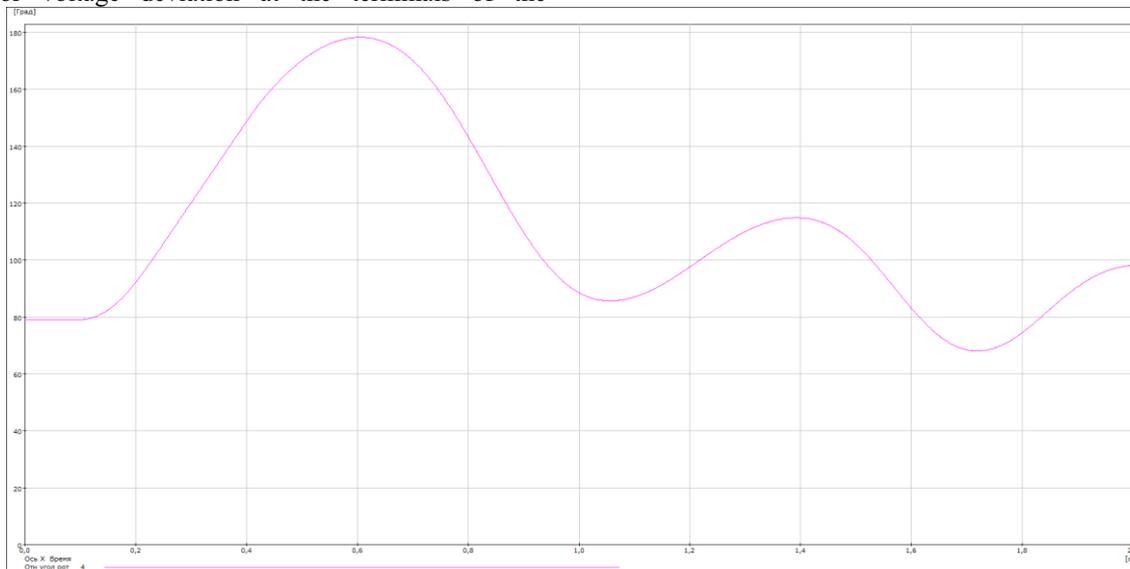


Fig.10. Rotor's angle of **G4** within 0.3 seconds **SC** (developed software package)

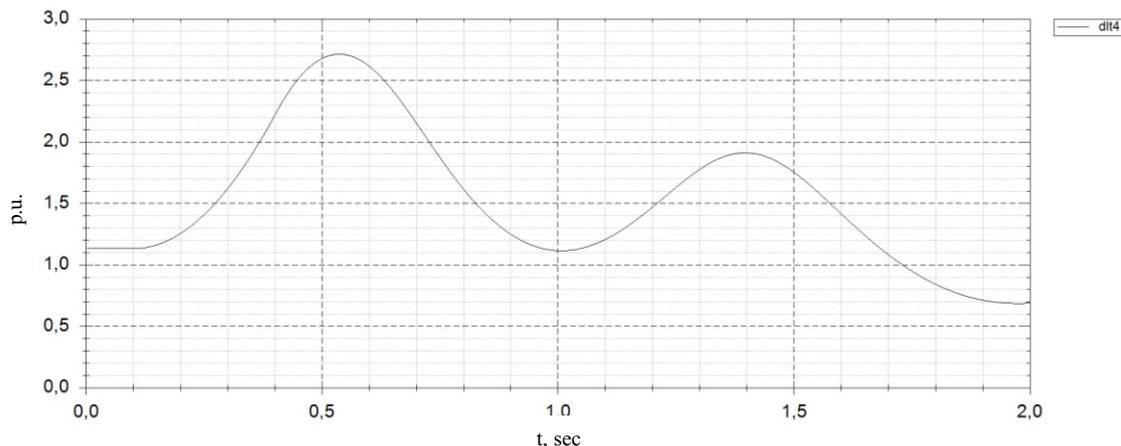


Fig.11. Rotor's angle of **G4** within 0.3 seconds **SC** (software package *Mustang*)

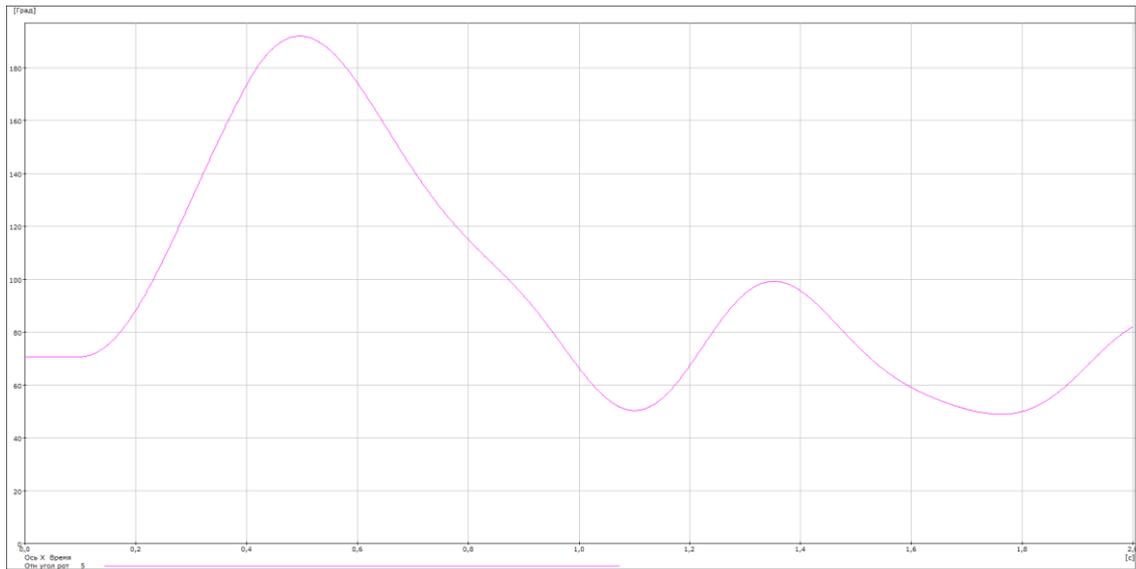


Fig.12. Rotor's angle of G5 within 0.3 seconds SC (software package Mustang)

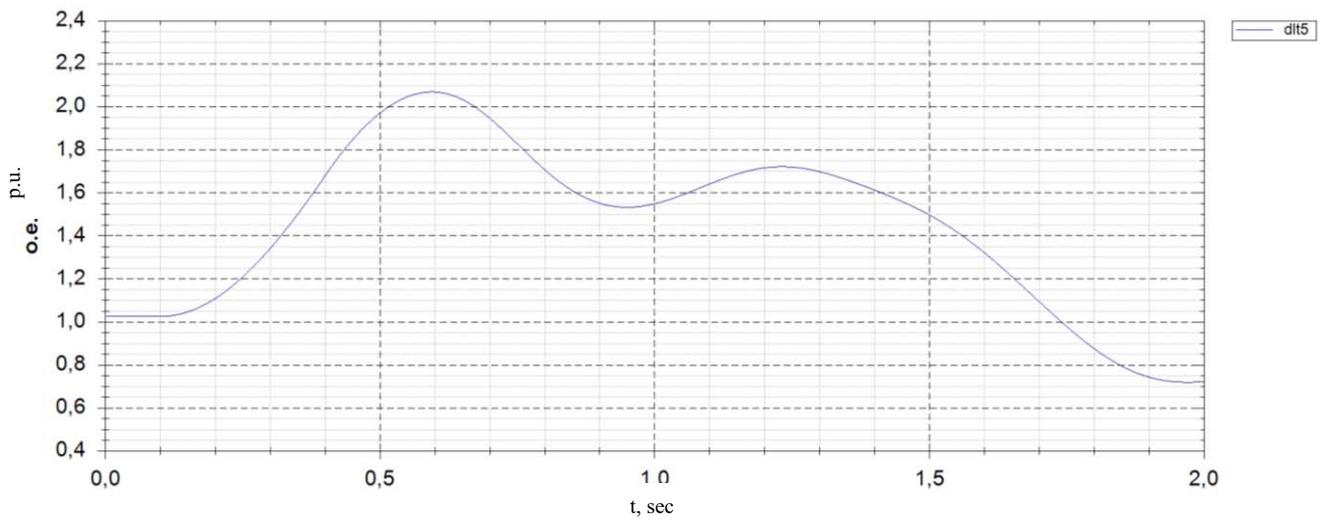


Fig.13. Rotor's angle of G5 within 0.3 seconds SC (developed software package)

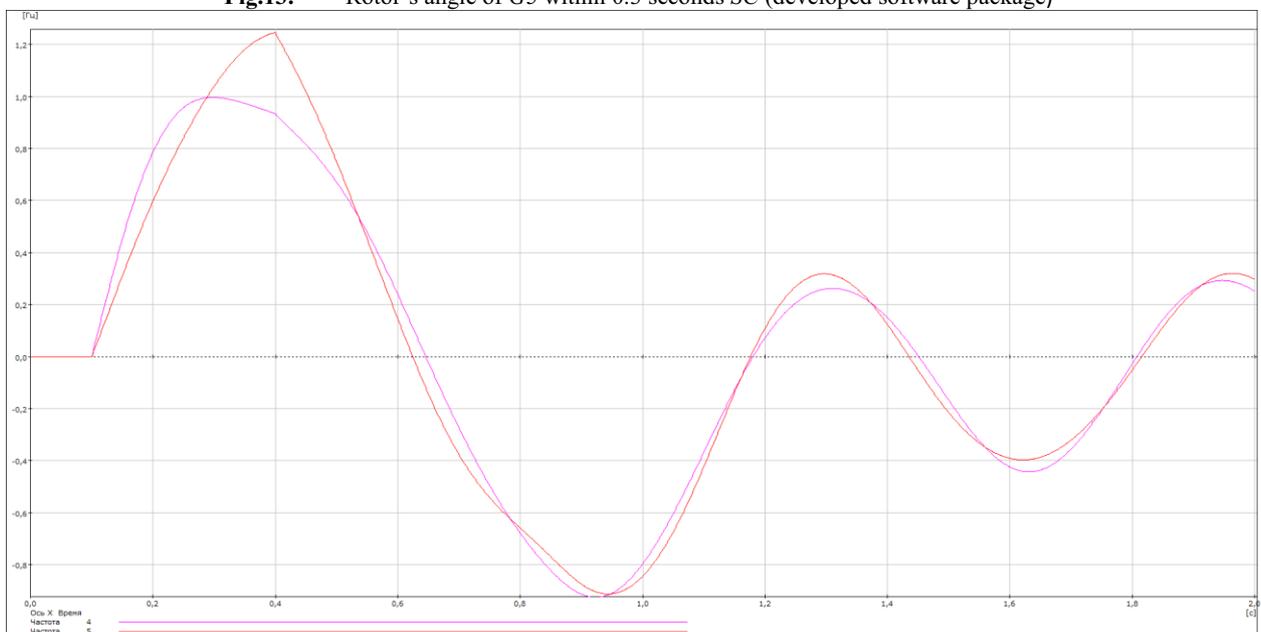


Fig.14. Rotor's angular frequency of G4 and G5 within 0.3 seconds SC (software package Mustang)

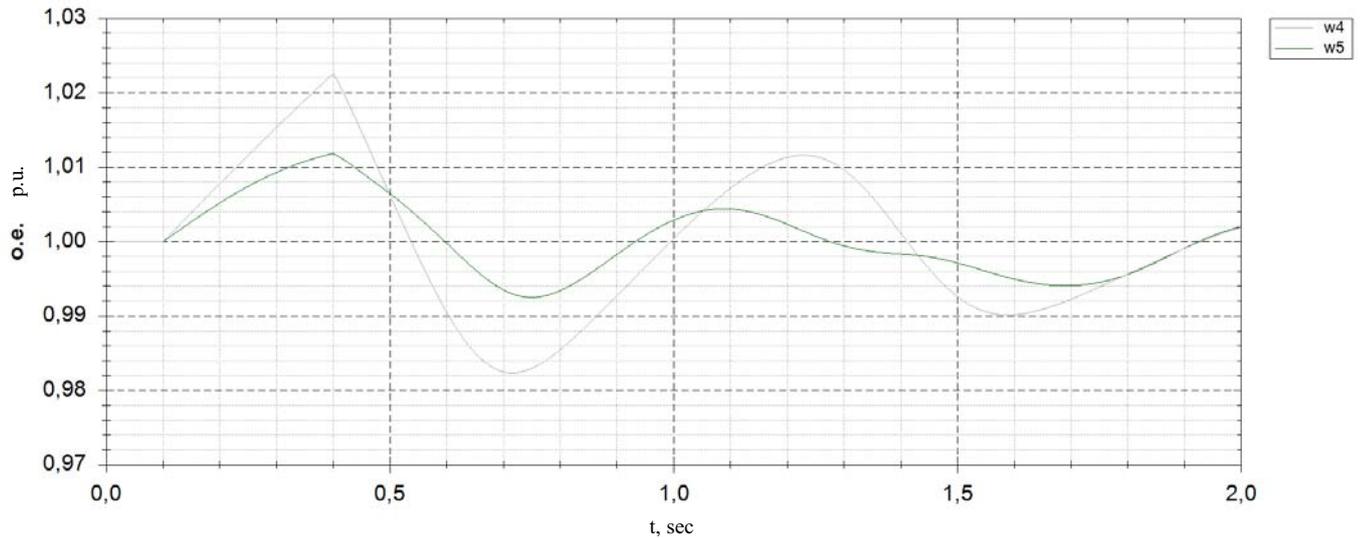


Fig.15. Rotor’s angular frequency of G4 and G5 within 0.3 seconds SC (developed software package)

The simulation results showed a high reliability of steady-state and transient modes calculations in the developed software package in comparison with similar calculations in the industrial software package.

4 Long-term dynamic processes in the grid with CPFC

When considering long transient processes, the power network diagram shown in Fig.7 was used. A three-phase SC on the 500 kV line was considered in the deficient part of the system (subsystem 1) in 0.3 seconds

at node 12. Comparing the transient process graphs shows a decrease in the disturbance impact in the deficient part of the grid, separated from a 220 kV grid by a combined device, on the surplus part of the PS. Thus, the influence of the disturbance in one part of the grid is damped by the combined device. The difference in the values characterizing the transient process, excluding and including CPFC, reaches 20-30%.

In the power system's deficient part, the CPFC installation can significantly reduce the generator rotors' swing amplitude. There is a damping of the mode parameters oscillations and faster damping of the transient process than in the scenario without CPFC.

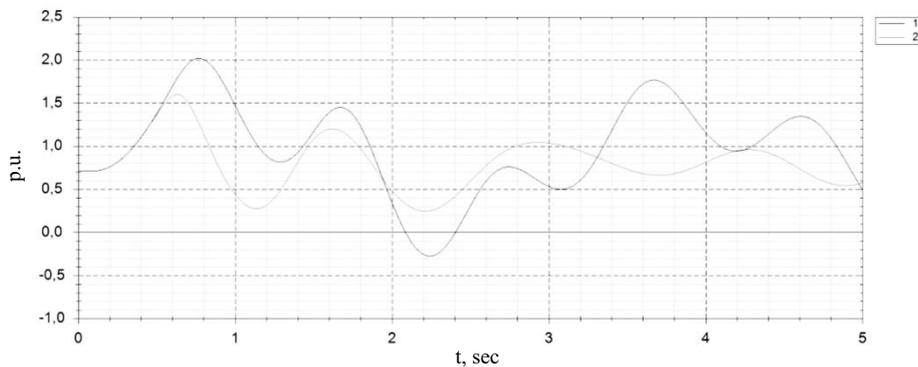


Fig.16. Rotor’s angle of G6:1 – without CPFC; 2 – with CPFC

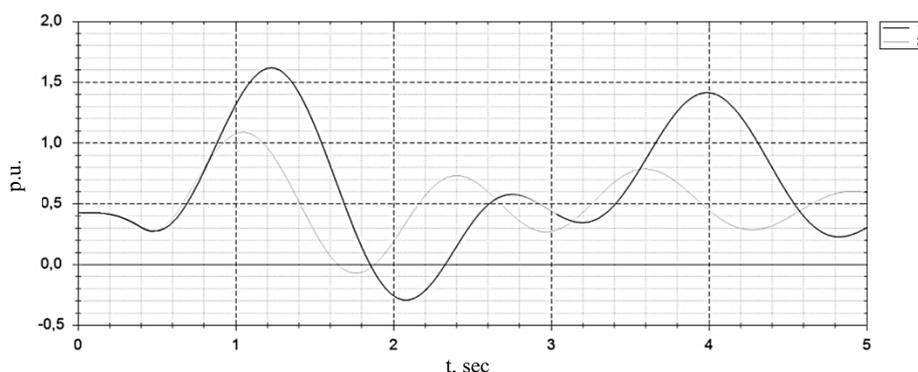


Fig.17. Rotor’s angle of G2: 1 – without CPFC; 2 – with CPFC

The analysis of the transient processes graphs excluding and including the **CPFC** installation showed that the installation could significantly reduce the generator's mutual oscillations in both subsystems. The combined device installation allows increasing the power generators' stability in the power system by damping the oscillations and reducing the time of the transient process.

At the emergency failure or maintenance switching of the power transmission line at the interconnection ends, the intermediate substations' load should not be disconnected. The **CPFC** carries out its power supply.

Thus, of particular interest is the power grid's operation with the installed combined device when the load is supplied directly from it. In this case, the **DFSMC** task is to maintain a given voltage level at its terminals and a certain value of the active power flow (for clarity, only the parameter functions are shown, the control law of which is changed in the algorithm, also the **AEC** links inertia is not demonstrated). The excitation voltage is calculated as follows:

$$E_{de1} = E_{de01} + k_{0\omega_1} \cdot (\omega_r - \omega_{r\ set1}) + k_{0\omega_1 dif} \cdot \omega_{dif}, \quad (24)$$

$$E_{de2} = E_{de02} + k_{0P} \cdot (P_g - P_{set}) + k_{0\omega_2 dif} \cdot \omega_{dif}, \quad (25)$$

where $\omega_r = s_r + 1$ and $\omega_{r\ set1} = (\omega_1 + \omega_2)/2$ are the actual and required rotation speed of the asynchronous electromechanical frequency converter shaft; $\omega_{dif} = \omega_1 - \omega_2$ is the frequency difference of the two subsystems; $k_{0\omega_1 dif}$ and $k_{0\omega_2 dif}$ are the gain factors for the frequency difference in subsystems 1, 2 (in the case under consideration, in the initial operation mode before the disturbance, the frequencies are equal, that is $\omega_1 = \omega_2$); $k_{0\omega_1}$, $k_{0\omega_2}$ and k_{0P} are the gain factors for the frequencies deviation in subsystems 1, 2 from the rated value and for the deviation of the actual transmitted power from the set value P_{set} , respectively; P_g is the actual value of the transmitted active power.

When the **SC** is eliminated, the power supply of a load part is provided only by the **CPFC**. After the **SC** elimination, the signal from the automatic line disconnection fixation is sent to the input of the **AEC** settings block of **DFSMC** machines. The fact that the line is disconnected leads to the shift of combined device operation mode from maintaining the given power flow to maintaining the rated frequency in the adjacent grid

$$E_{de1} = E_{de01} + k_{0\omega_1}(\omega_{r1} - \omega_{set1}) + k_{0\omega_1 dif} \cdot \omega_{dif} + k_{\omega_1}^D \frac{d\Delta\omega_{r1}}{dt}, \quad (26)$$

$$E_{de2} = E_{de02} + k_{0\omega_2}(\omega_{r2} - \omega_{set2}) + k_{0\omega_2 dif} \cdot \omega_{dif} + k_{\omega_2}^D \frac{d\Delta\omega_{r2}}{dt}, \quad (27)$$

where $k_{\omega_1}^D$ and $k_{\omega_2}^D$ are the gain factors of the derivative of the **DFSMC** shaft rotation speed change for subsystems 1 and 2.

A diagram of the grid under consideration with a detailed layout of the **DFSMC** connection point is presented in Fig.18.

When the **AT** is taken out for maintenance, the transmission of electric power by the combined device is carried out only by **DFSMC** machines.

The **SC** was investigated on line 6-7 with its subsequent outage. At the same time, in the initial operation mode, the **M1** machine, connected to node 6 of the scheme, operates in a motor mode, that is, it consumes active power from the grid. The **M2** machine, in its turn, operates in a generator mode and provides active power to node 5.

As a result of the **SC** clearance on the line 6-7, the local load in node 6 is not disconnected and its power supply is carried out by the **DFSMC**. In this case, the **DFSMC** is automatically reversed: the **M1** machine goes into the generator mode and the **M2** – into the motor mode (Fig.19).

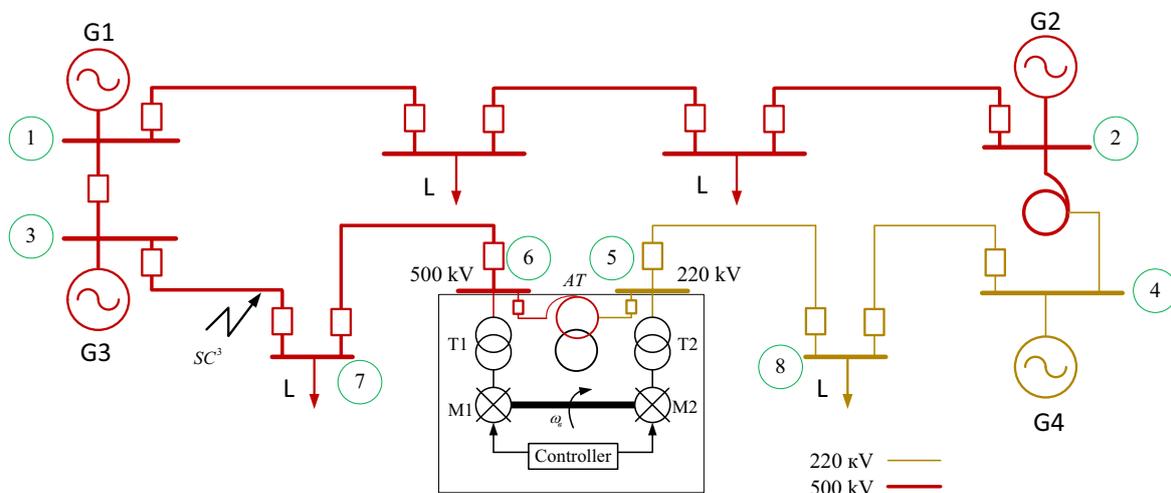


Fig.18. Principal scheme of the grid

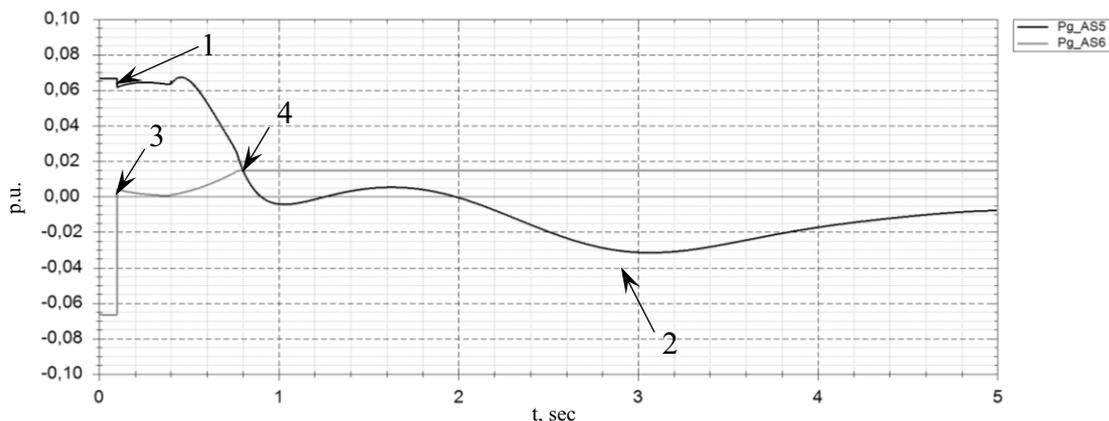


Fig.19. Active power of $M1$ and $M2$ within the long-term dynamic process

According to the calculations, the reverse of the **DFSMC** power is carried out in no more than 3 seconds, while the power supply to consumers from the **M1** machine is resumed in 0.4 seconds due to the asynchronous machine's rotors kinetic energy. It should be noted that the frequency in the adjacent grid does not change significantly and does not go beyond the operation settings of the existing automation devices.

Thus, the performed calculations proved the high efficiency of using the combined device in terms of improving the transient electromechanical processes quality, and also made it possible to form an algorithm for controlling **DFSMC** machines in the case of its isolation to a balanced power region without deteriorating the power supply quality to consumers.

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