

Design and simulation of current-fed dual-active full-bridge DC/DC converter control system applied to proton exchange membrane fuel cell

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Abstract. Proton exchange membrane fuel cells (PEMFC) have been increasingly applied in clean and efficient distributed power generation systems in recent years. However, the output characteristics of PEMFC are relatively soft due to the influence of multiple polarization overvoltage. With the increase of current density, the output voltage shows a nonlinear downward trend. In addition, in order to ensure the lifetime of PEMFC, the output ripple current is often concerned. Considering the above characteristics, in order to achieve efficient power transmission of the PEMFC system, this paper adopts the current-fed dual-active full-bridge (DAB) DC/DC converter topology, which can achieve soft switching of most switches. Furthermore, it can restrain the characteristics of the current ripple. Hence, the power transmission efficiency and current ripple requirements of PEMFC are fulfilled. Finally, the current-fed DAB is designed and simulated through PSIM. The C-block is applied to simulate the PEMFC nonlinear power supply according to the actual PEMFC parameters. In order to achieve efficient constant power transmission of the nonlinear PEMFC, a novel power closed-loop control strategy is proposed.

Keywords: Current-fed DAB; Proton exchange membrane fuel cell; Soft switch; DC/DC converter.

1 Introduction

As a new power generation technology, the hydrogen fuel cell energy's conversion is not restricted to Carnot cycle, with features of high energy conversion rate^[1], clean without pollution, and zero noise, which provides a new route for the solution of energy crisis and environmental pollution. PEMFC can take full advantage of new energy like industrial by-product hydrogen and abandoned wind and light to produce hydrogen to achieve decoupling of hydrogen and new energy development and utilization in different time scale, which can effectively achieve demand side management and with important roles to play in areas like improving energy comprehensive utilization level, reducing energy use cost, promoting distributive power connection, guarantee power supply and demand response.

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Due to the soft feature and big fluctuation of the output voltage generated by PEMFC, as well as the slow response of output power and only one-way flow, a converter must be equipped to regulate, control and manage the output of fuel cell before grid connection operation. In the paper, a converter topology is adopted to PEMFC whose voltage change range of 40V~75V as input power, which connects to 750V DC bus voltage end and transmit at constant power.

Due to the soft output feature of PEMFC, it is easy to be affected by load, and its output power is unstable under no closed loop control, meanwhile in order to ensure PEMFC's life, it needs to adopt a DC/DC converter with high step-up ratio, high efficiency, and small input current ripple wave. Because of its features of achieving high power density, natural zero voltage switch (ZVS) characteristics, high efficiency, symmetrical structure and simple control, a dual-active bridge bidirectional DC/DC converter (Dual-Active-Bridge, DAB) is widely used in medium, and large power supplies. However, voltage source DAB has some disadvantages, such as large input ripple current, limited soft switching range, large circulating current through devices and transformers, and low efficiency in high voltage and high current applications. In [2], a current type DAB converter is applied into the photovoltaic system that needs to have high step-up ratio DC/DC converter. In order to increase gain of voltage conversion, the current-fed type DC/DC converter is the best choice^[3-6]. Current type DAB has been proved to be applied to low voltage and big current field.

In [7], a current-fed type dual-active full-bridge DC/DC converter is proposed, where that converter eliminates the clamp forward circuit and passive buffer device in the circuit by secondary modulation technique, and realizes zero current shutdown (ZCS) and zero voltage switching of the primary side device and zero voltage switching of secondary side device, which significantly reduces switching loss.

In this paper, a PEMFC's structure and external features are provided, and the working principles of that current type DAB topological structure then are analyzed. According to the proposed new power closed loop control strategy, a power closed loop design and simulation modeling of non-linear PEMFC voltage source is achieved, so as to achieve the effective constant power transmission.

2 PEMFC structure and external features

2.1 PEMFC structure

PEMFC is composed of cathode tank, anode tank and electrolyte membrane (proton exchange membrane). Hydrogen reaches to anode through pipes or gas guide plate. Under the action of anode catalyst, hydrogen molecules dissociate into positively charged hydrogen ions (protons) and release negatively charged electrons. Hydrogen ions travel through the electrolyte (proton exchange membrane) to the cathode, while electrons travel through an external circuit to the cathode. Electrons form an electric current in an external circuit, and with appropriate connections, it can output electrical energy to a load. At the other end of the cell, oxygen (or air) travels to the cathode through a pipe or gas guide plate. Under the action of cathode catalyst, oxygen reacts with hydrogen ions and electrons to produce water, meanwhile releasing heat.

2.2 PEMFC external features

Due to the special loss mechanism such as activation overvoltage, ohm overvoltage and concentration overvoltage, in operation, the single PEMFC voltage drops occur as the

output current changes. Under the medium power level, PEMFC STACK's output voltage composed of multiple PEMFCs generates big change with the load fluctuation. With the increase of PEMFC STACK, the output current increases continuously and maintains at a relatively high level. On the contrary, the output voltage is often relatively low. So as to the application occasions where voltage is low and current is big of PEMFC generation system, the DC/DC converter must be matched. Only in this way, the output voltage can be kept at a relatively high level.

Referring to the output characteristic curve of Nedstack manufacture's model FCS 10-XXL's PEMFC, as shown in Fig.1, the output voltage reduces with the current increase, and its reduction scale is big. So this paper combines the current type DAB with high voltage gain for PEMFC, so as to achieve 750V DC bus constant power transmission.

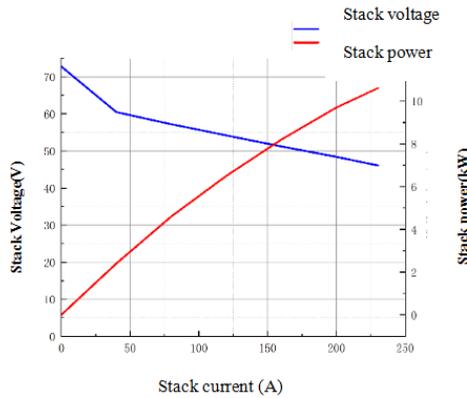


Fig. 1. PEMFC output characteristic curve.

3 PEMFC current-fed type dual-active full-bridge DC/DC converter

3.1 Topological structure and features

PEMFC current-fed type dual-active full-bridge DC/DC converter (DAB) topology is shown as Fig.2. D_0 is the reverse blocking diode, which prevents the negative current flows into battery pack and cause damage to PEMFC. CF-DAB is composed of two H bridges and one high frequency transformer, altogether 8 switch tubes, $S_1 \sim S_8$ respectively. Its inverse parallel diodes are $D_1 \sim D_8$ respectively. The buffer capacitance of each switch tube is $C_1 \sim C_8$, L is boost inductor, L_{lk} is transformer leakage inductance, and C_0 is output capacitance.

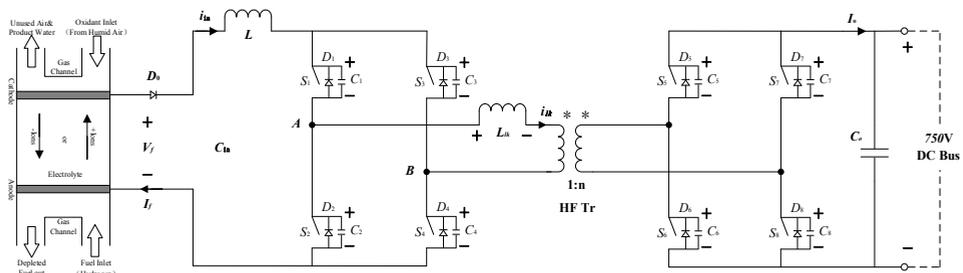


Fig. 2. PEMFC current-fed dual-active full-bridge DC/DC converter.

By matching of duty cycle of the primary and secondary side switch tube, this converter achieves the zero current shutdown and zero voltage switching of primary side switch tube, and the zero voltage shutdown of the secondary side switch tube, which reduces the switching loss of switch tubes. Due to the voltage reflection of DC power grid, the voltage of the primary side switch tube is clamped at V_o/n , which eliminates voltage spike. The converter transmission efficiency is greatly improved in this circuit.

3.2 Operating model analysis

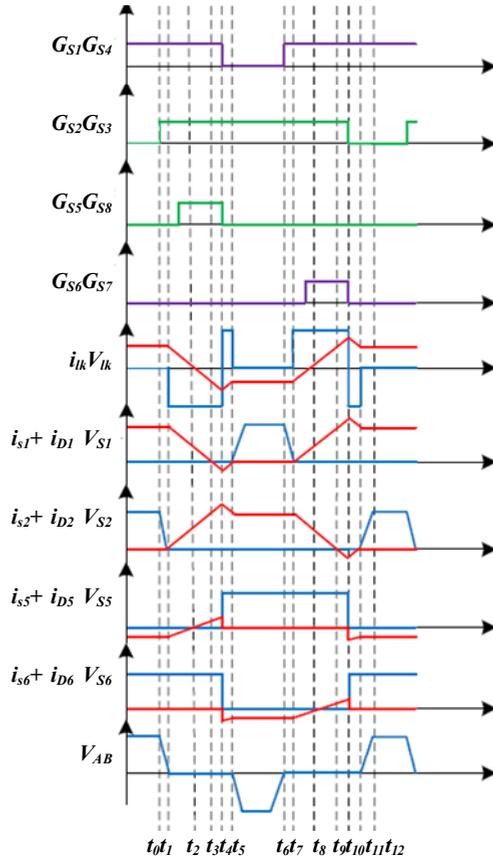


Fig. 3. Working waveform.

Working waveform is shown as Fig.3. Before t_0 , the primary side switch tubes S_1, S_4 and the secondary side diodes D_5, D_8 turn on, the power transmits from primary side to secondary side. $i_{lk} = i_{s1} = i_{s4} = i_{in}$, $i_{D5} = i_{D8} = i_{in}/n$.

Operating model 1 [t_0, t_1]: In t_0-t_1 period, the switch tubes S_2 and S_3 are triggered, and then buffer capacitance C_2 and C_3 are discharged. Because of the extremely low buffer capacitance and extremely short time, power transmits from primary side to secondary side. At the end of this period, the voltage of switch tubes S_2, S_3 reduces to zero, and zero voltage start is enforced.

Operating model 2 [t_1, t_2]: In t_1-t_2 period, S_2 and S_3 are turned on, all of switch tubes in the same bridge arms of primary side keep conducting state, the voltage of the primary side leakage inductance is V_o/n , at this time $V_{lk} = V_o/n$. The leakage inductance current starts to decrease linearly with slope k . i_{s2} and i_{s3} begin to increase from zero with $-k/2$ slope, i_{s1} and

i_{s4} reduce linearly with $k/2$ slope, power stores on leakage inductance L_{lk} begins to transmit to secondary side. At the end of this period, S_5 and S_8 begins are triggered. $i_{lk}=i_{D5}=i_{D8}=0$, the slope is $k=V_o/nL_{lk}$. Owing to the D_5 and D_8 turning on, the voltage of switch tubes S_5 and S_8 are clamped to zero before turning on.

Operating model 3 [t_2, t_3]: At the moment of $t=t_2$, for the switch tubes S_5 and S_8 , ZVS is completed. The voltage reflected from secondary side to the primary side is still V_o/n , so the leakage inductance current i_{lk} continues to reduce with the same slope k . Leakage inductance is to absorb power, and secondary side power feeds back to primary side. i_{s2} and i_{s3} continue to increase with $-k/2$ slope, and the currents of the i_{s1} and i_{s4} continue to reduce linearly with $k/2$ slope. At the end of this period, $i_{s1}=i_{s4}=0$, $i_{s2}=i_{s3}=-i_{in}$, the two currents i_{D1} and i_{D4} begin to stream conducting, making S_1 and S_4 voltage clamped at zero and maintain in t_3-t_5 period.

Operating model 4 [t_3, t_4]: In t_3-t_4 period, Two switches of i_{D1} and i_{D4} are opened continuously, the status of the secondary side switch is unchanged, leakage inductance current i_{lk} continues to reduce with slope k , i_{s2} and i_{s3} continue to increase with $-k/2$ slope, and leakage inductance continues to absorb power. At the end of this period, S_5 and S_8 stop to be triggered, S_5 and S_8 current transfer to D_6 and D_7 . The polarity of the reflected voltage on the leakage inductance is opposite, and the absolute value is unchanged. Leakage inductance current i_{lk} begins to change at the opposite direction with slope $-k$, all of the leakage inductance current i_{lk} and i_{s2} , i_{s3} reach peak value.

Operating model 5 [t_4, t_5]: In t_4-t_5 period, D_6 and D_7 stream conduct, the polarity of the reflected voltage on the leakage inductance is opposite, and the absolute value is unchanged. Leakage inductance current i_{lk} begins to increase from peak value with slope $-k$, the two currents of i_{s2} and i_{s3} begin to reduce from peak value with slope $-k/2$, and leakage inductance releases power to the secondary side. At the end of this period, $i_{lk}=-i_{in}$, $i_{D6}=i_{D7}=-i_{in}/n$, $i_{s2}=i_{s3}=i_{in}$, $i_{D1}=i_{D4}=0$. Buffer resistance C_1 and C_4 begin to charge, switch tubes S_1 and S_4 achieve ZCS, which reduce turn-off loss.

Operating model 6 [t_5, t_6]: In t_5-t_6 period, the power is transferred from primary side to secondary side, the fuel cell charges to buffer capacitance C_1 and C_4 and its value is V_o/n .

Operating model 7 [t_6, t_7]: In t_6-t_7 period, the converter achieves a convert of fuel cell output current from switch tubes S_1, S_4 to switch tubes S_2, S_3 , the circuit enters a stable condition, the power is transferred from primary side to secondary side, $i_{lk}=i_{s2}=i_{s3}=-i_{in}$. It can be known from the analysis process from operating model 1 to operating model 6, in which this switch process realizes soft switch and eliminates voltage spike.

Operating model 1 to 7 describes the power transmission in half the cycle and the process of realizing soft switch. The operating model curves are symmetrical. In the next half cycle, the process is similar, and it will not be described further here.

3.3 Switch loss calculation

It can be known from the power transmission process of half the cycle that S_2, S_3 and S_5, S_8 realize ZVS turning on, S_1 and S_4 realize ZCS turning off. At the moment of $t=t_4$, S_5 and S_8 are not triggered, its current directional is transferred. During the transfer process, S_5 and S_8 are hard turn off, and D_6 and D_7 realize hard turning on. The half cycle power transmission loss can be obtained by the calculation of Formula (1):

$$P_{loss} = 2 \times \frac{1}{T} \left[\int_{t_5}^{t_5+t_{off}} i_{s5}(t) \times V_{s5}(t) \times f_{sw} dt + \int_{t_5}^{t_5+t_{on}} i_{D6}(t) \times V_{D6}(t) \times f_{sw} dt \right] \quad (1)$$

Where T is switching period, f_{sw} is switch frequency, t_{off} is S_5, S_8 turn-off time, t_{on} is D_6, D_7 turn-on time. $i_{s5}(t)$, $V_{s5}(t)$ are S_5 's voltage and current expression formulas in $t_5-t_5+t_{off}$

period respectively, $i_{D6}(t)$ and $V_{D6}(t)$ are D_6 voltage and current expression formulas in $t_5 \sim t_5 + t_{on}$ period respectively. It can be know from topological structure symmetry that the entire cycle loss is twice P_{loss} .

4 Circuit parameters and controller design

4.1 Main circuit parameter calculations

In order to achieve the design of DC/DC converter for PEMFC, each model source should keep constantly 7.5kW output power. As the output voltage range of fuel cell is from 40V to 75V with big change scope, the choice of transformer turns ratio n is considered by the actual needs of high step-up ratio, as well as to consider the impact of turns ratio n on transmission loss. Therefore, available turns ratio should be selected to get lower switch loss and higher transmission efficiency. So lay down $n=7$, duty cycle $D=0.75$, switch frequency $f_s=10kHz$.

According to Formula (2) transformer's leakage inductance is:

$$L_{lk} = \frac{V_o \cdot (D - 0.5)}{2n \cdot I_{in} \cdot f_s} \quad (2)$$

From turns ratio n , leakage inductance L_{lk} and V_o determines the changing speed of leakage inductance current i_{lk} is determined. And this value is fixed. The corresponding model duration can be calculated according to current slope. The specific results are obtained from (3) calculation:

$$\begin{cases} \Delta i_{lk} = 0 - i_{in} \\ \frac{\Delta i_{lk}}{\Delta t_{12}} = \frac{di_{lk}}{dt} \\ \frac{di_{lk}}{dt} = -\frac{V_o}{nL_{lk}} \end{cases} \quad (3)$$

Where Δt_{12} is the specific time of operating model 2, Δi_{lk} is the changing amount of leakage inductance current at that time period. Boost inductance L ensures current's stable and continuous. In case of closed loop, the boost inductance can be obtained from Formula (4) [7] is:

$$L \geq \frac{V_{in} \cdot (D - 0.5)}{\Delta i_{in} \cdot f_s} \quad (4)$$

Where Δi_{in} is the ripple wave current through L .

4.2 Power closed loop design

Two polarization effects of activation polarization and ohm polarization are considered in this paper. The corresponding polarization curves were fitted by actual PEMFC parameters. In PSIM by using C-Block module the non-linear PEMFC external feature model was built. By testing the primary side power P_{in} of current-fed DC/DC converter and compare it to

P_{ref} , and the difference e_{rr} is input into PI controller, the output power was clipped to obtain duty cycle d' to adjust t_{34} and t_{67} . According to current type DAB wave generating law, output duty ratio d is used to adjust duty ratio of different switch tubes, eventually the output power is kept a stable value through controlling the size of transmission power. The closed loop system control frame figure is shown as Fig.4. In order to simulate big capacity 750V DC bus, increase the magnitude of DC bus capacitance C_0 in PSIM. Meanwhile to ensure 7.5kW constant power output, take R_0 as 75Ω , so there's method to close loop control the output side power. As the power transmission efficiency of current type DAB, the difference between input and output sides is not so big, hence the output power P_{in} of the non-linear power PEMFC is taken as the control objective. This similarity is only used for simulated analysis, in actual circuit, the output power is still used to conduct closed loop control.

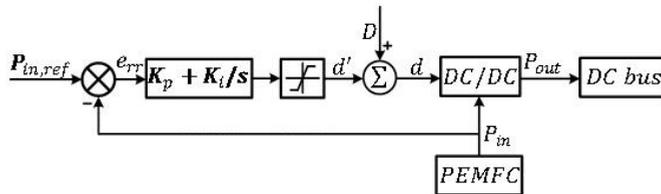


Fig. 4. Control block diagram of closed loop system.

5 PSIM emulation

The non-linear PEMFC module is built by using PSIM, and the main circuit of current-fed dual-active full-bridge DC/DC converter is shown as Fig.5. The main circuit parameters are set according to parameter calculation in section 4.1. By collecting the output voltage and current signals of PEMFC, it is fed back to the power closed-loop algorithm module for calculation, and the calculation results are input into the current type DAB modulation module to adjust the duty cycle of different switches according to the wave generation law of current type DAB.

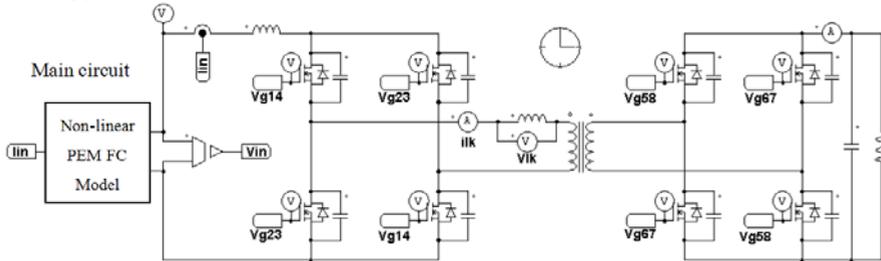


Fig. 5. PEMFC current-fed dual-active full-bridge DC/DC converter PSIM main circuit structure diagram.

Fig.6 shows the actual voltage and current oscillogram of non-linear PEMFC module. Under the effect of power closed loop, during activation polarization and ohm polarization, the output voltage is reduced with the current increasing, and eventually the voltage and current tend to be stabilized.

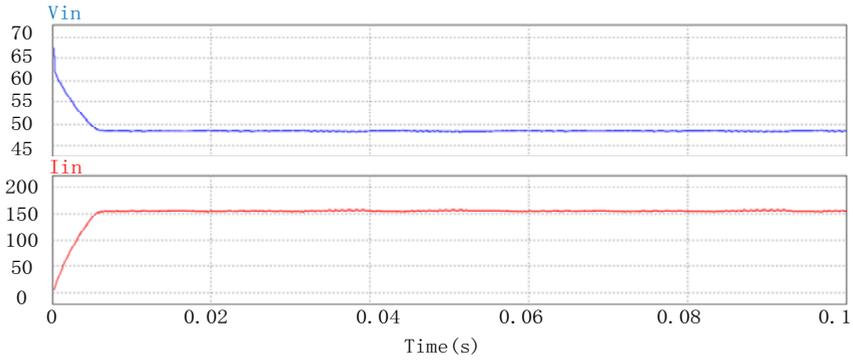


Fig. 6. PEMFC non-linear working waveform.

Fig.7 is the waveform of leakage inductance L_{lk} voltage and current. During the entire dynamic process, the voltage does not have obvious peak, and the current has always been appearing dynamic change. Within half of the period, the occurrence of i_{lk} current peak makes diodes D_1 and D_4 stream conduct, and ensure S_1 and S_4 zero current shutdown.

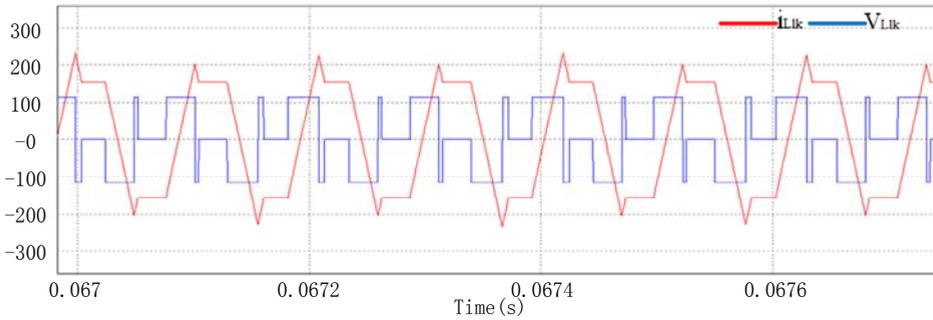


Fig. 7. Leakage inductance voltage and current waveform.

Fig.8 shows the realization process of soft switch, and only one switch tube with the same PWM control signals is displayed. For example, S_1 and S_4 have the same control signals, S_1 voltage and current waveform is only shown in the Fig.9. In one cycle, S_5 and S_6 are ZVS, and S_1 and S_2 are ZVS and ZCS. Similarly in the same cycle, switch tubes S_2, S_3 and S_6, S_7 staggering corresponding with the same bridge S_1, S_2 and S_5, S_8 can also correspondingly realize soft switch operating.

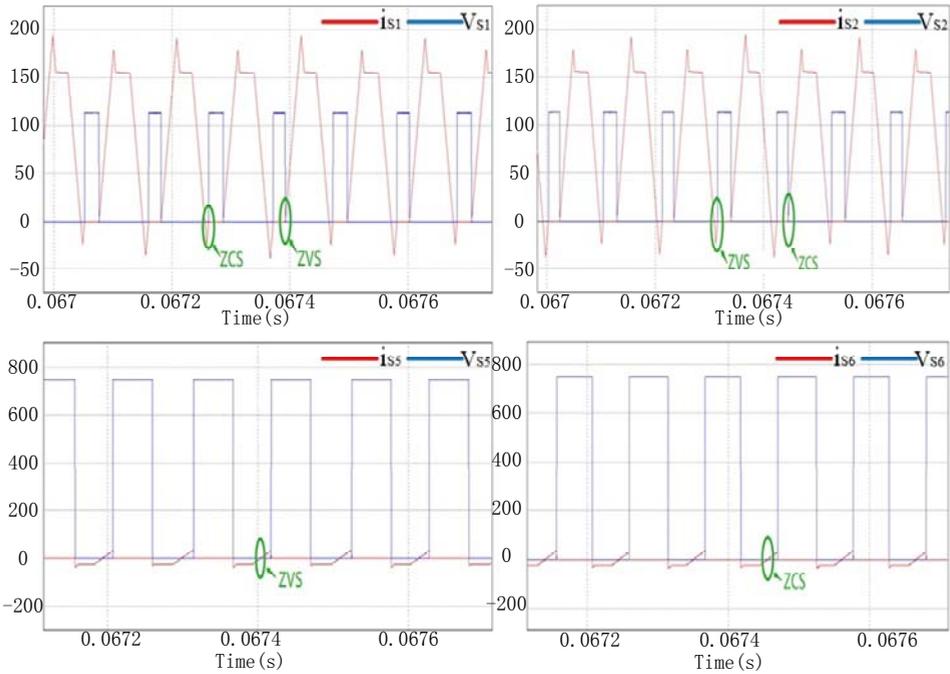


Fig. 8. Diagram of main waveforms of DC/DC converte.

In Fig.9, input side current ripple wave of current type DAB is shown. the size of the input current ripple wave covers 0.478% of the current average value. According to related documents conclusion, the high frequency current ripple wave (1kHz, 10% ripple wave current amplitude) will not pose impact on PEMFC’s service life^[9], thus satisfying the design requirement of the output current ripple wave.

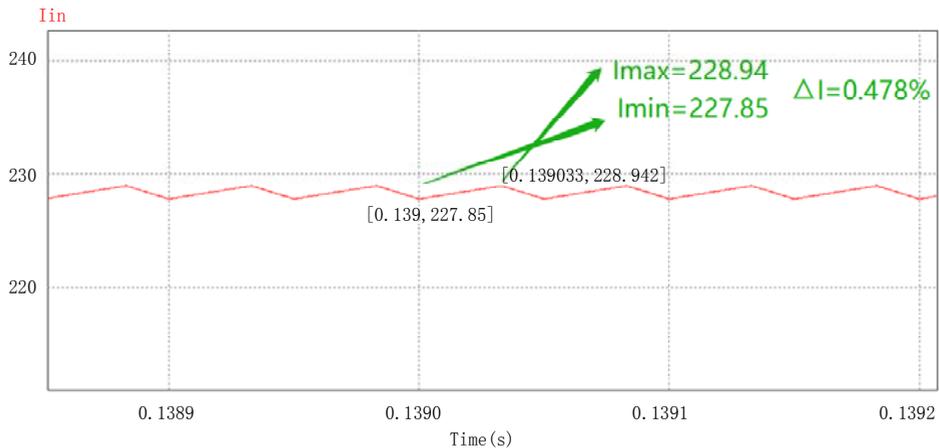


Fig. 9. DC/DC converter input current ripple expansion diagram.

Fig.10 is the input/output power waveform diagram under closed loop control. 750V DC power grid side needs constant 7.5kW power, but during the time period of 0 to 7.5kW, the supply and demand is imbalanced for the output power of the non-linear PEMFC output power. Because the DC power grid is simulated by capacitor, the capacitor outputs 7.5kW

constant power to the resistor R_0 , the capacitor voltage drops briefly, until the end of this transition process, the input and output power tends to be stable.

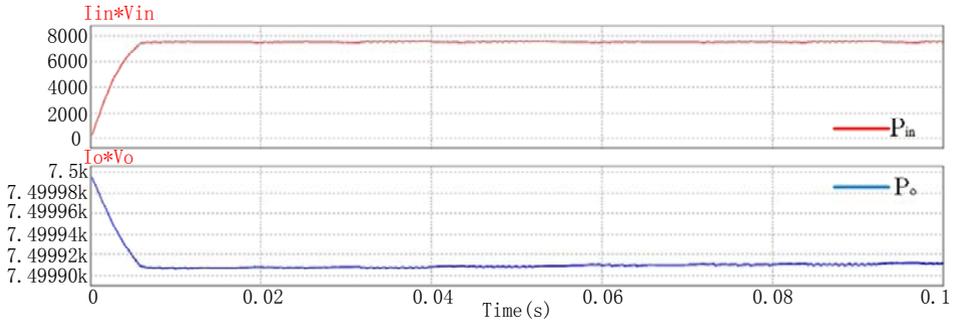


Fig. 10. DC/DC converter input and output power waveform.

In Fig.11, the expansion waveform of input and output power is shown, which is the average power of the input and output power in the displayed zone. The input average power is 7.52kW, output power is 7.47kW, the efficiency is 98.57%.

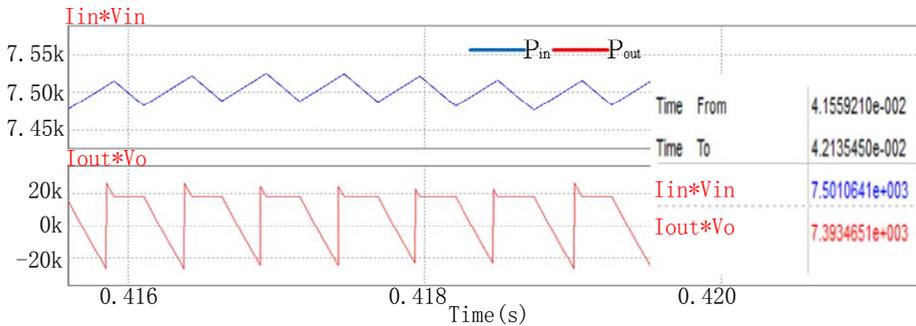


Fig. 11. DC/DC converter input and output power waveform expansion diagram.

6 Conclusion

In this paper, considering the nonlinear output characteristics of PEMFC and the high requirement of output current ripple, the current model DAB is applied to the constant power output of PEMFC to 750V DC bus. Through theoretical analysis, power closed-loop design and simulation experiments, it is verified that the topology can achieve efficient and constant power transmission. In addition, the current fed dual active full bridge DC/DC converter achieves zero voltage on, zero current off of primary switches and zero voltage on of secondary switches, and 0.478% high frequency input current ripple will not affect the life of PEMFC, which meets the requirements of constant power transmission efficiency and current ripple of nonlinear power supply PEMFC.

Acknowledgments

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